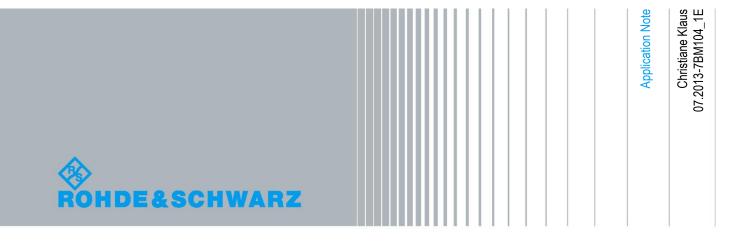
# DAB Transmitter Measurements for Acceptance, Commissioning and Maintenance Application Note

#### **Product:**

| R&S<sup>®</sup>ETL

Broadcasting transmitters are subject to particularly stringent standards with respect to broadcast signal quality, because even small faults can lead to service disruptions for many listeners.

A single instrument, the R&S<sup>®</sup>ETL TV analyzer, performs all required DAB transmitter measurements, from the initial acceptance testing for the transmitter, to measurements performed during commissioning and preventive maintenance.



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## 1 Overview

Broadcasting transmitters are subject to particularly stringent standards with respect to broadcast signal quality, because even small faults can lead to service disruptions for many listeners.

A single instrument, the R&S<sup>®</sup>ETL TV analyzer, performs all required DAB transmitter measurements, from the initial acceptance testing for the transmitter, to measurements performed during commissioning and preventive maintenance.

The measurements described here satisfy many country-specific and customerspecific test specifications. Users need only set the limit values accordingly.

Section 2 describes the preparatory steps. These include the necessary test equipment and setup, as well as steps to protect the T&M equipment against destructively high input power. This is followed by a description of typical default configurations for the R&S<sup>®</sup>ETL.

Section 3 lists the various measurements. For every reserve system in the transmitter, these measurements should be repeated at least once during acceptance testing. Maintenance measurements, on the other hand, can initially be limited to power, MER and BER, and then expanded only as needed.

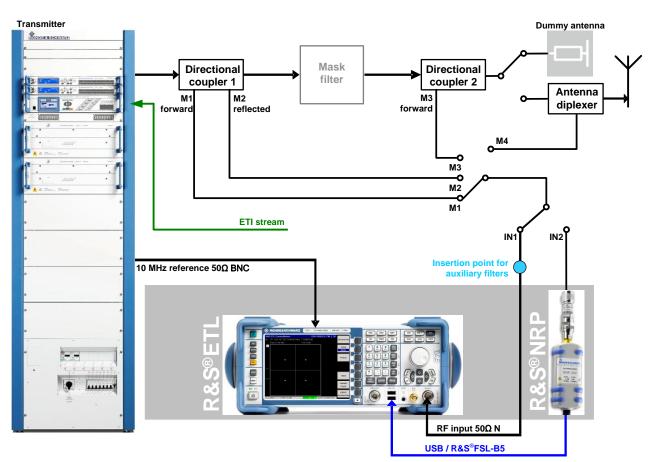
Additional background information on this topic can be found in the book "Digital Video and Audio Broadcasting Technology" by Walter Fischer [1].

## 2 Preparatory Steps

## 2.1 Required Equipment

Basic configuration	
	<ul> <li>R&amp;S<sup>®</sup>ETL TV analyzer with:</li> <li>options as needed (see Section 7)</li> <li>current firmware (available at no cost at <u>www.rohde-schwarz.com/product/ETL.html</u>)</li> </ul>
Application- or measu	urement-specific configurations
	Transmitter operation without signal broadcasting for transmitter ac- ceptance testing or commissioning
	Dummy antenna
ETI	For transmitter acceptance testing
Stream Generator	ETI stream generator
	For Transmitter Output Level measurements (3.1.1) with an inaccuracy of < 0.1 dB
	Additional power sensor, e.g. R&S <sup>®</sup> NRP-Z91
	For Shoulder Attenuation and Adjacent Channel Emissions measurements (3.3.1) using the variant "after mask filter"
$\mathcal{K}$	Notch filter to attenuate the wanted signal by 40 dB or more
	For Harmonics measurements (3.3.2)
$\approx$	Highpass filter with at least 40 dB wanted signal attenuation or more
	For Harmonics measurements (3.3.2) in the L-band
	Spectrum analyzer up to 6 GHz, e.g. R&S <sup>®</sup> FSL6

Test Setup



### 2.2 Test Setup

#### Fig. 1: Setup

DAB uses the ensemble transport interface (ETI) stream to implement synchronous data transmission. A DAB-compliant ETI stream is applied to the DAB modulator. Alternatively, some modulators can also generate a PRBS signal; however, the modulator input is not tested in this situation. The ETI feed present at the transmitter location is used for commissioning and maintenance measurements.

For transmitter acceptance measurements, the transmitter output is connected to a dummy antenna.

During commissioning, the measurements are initially performed using a dummy antenna, before the broadcast signal is applied to the transmit antenna via the antenna combiner. As a result, the test port at the antenna combiner (M4) is available as an additional measurement point.

During maintenance measurements, the signal is applied to the antenna combiner for broadcasting.

The RF input of the R&S<sup>®</sup>ETL (IN1) or the optional power sensor (IN2) is connected as follows for the various measurements:

- to the test port on the transmitter output (M1=forward, M2=reverse)
- to the test port behind the mask filter (M3)

If installed, the mask filter is located between the transmitter output and the dummy antenna or the antenna combiner. Some measurements can be taken at the test port before or after the mask filter (M1 / M3). The port to be used depends on which ports are available and which influencing factors are to be measured.

Some out-of-band emission measurements (see 3.3) require auxiliary filters, such as an adjustable notch filter. If they are required, these filters are added at the insertion point for auxiliary filters.

The EXT REF reference input located at the rear of the R&S<sup>®</sup>ETL TV analyzer is used to connect the instrument to the 10 MHz GPS time reference available at the transmitter station. The optional power sensor can be connected to the R&S<sup>®</sup>ETL via USB or via the sensor input on the R&S<sup>®</sup>ETL hardware option R&S<sup>®</sup>FSL-B5.

### 2.3 Protection Against Destructive Input Power

The R&S<sup>®</sup>ETL allows maximum input power peaks of 36 dBm (short-term, < 3 s), while the recommended, separate R&S<sup>®</sup>NRP-Z91 power sensor can handle up to 23 dBm.

It is therefore recommended that additional attenuators be used as needed to limit the average total power at the individual test ports to a range from 0 dBm to 10 dBm. This range provides adequate protection against short-term power peaks, while having a negligible effect on the instrument accuracy. The resulting attenuation must of course be taken into consideration during the measurements.

**R&S®ETL** Default Configuration

FREQ

### 2.4 R&S<sup>®</sup>ETL Default Configuration

The following conventions are used in these procedures:

- Terms in all caps refer to key labels, e.g. "FREQ" for L
- All other terms refer to the softkeys that are currently displayed along the righthand side of the screen. Arrows (→) separate the keys to be pressed in sequence

The following default settings apply to the R&S<sup>®</sup>ETL unless explicitly stated otherwise:

Spectrum analyzer mode
SETUP→Reference Ext: Use the external 10 MHz reference frequency
MODE→Spectrum Analyzer
FREQ→Center: Set to center frequency at mid-channel
SPAN→Span Manual: Set to 5 MHz
TRACE→Detector Manual Select→Detector RMS
BW→Res BW Manual: Set to 3 kHz
SWEEP→Sweeptime Manual: Set to 2 s
$AMPT \rightarrow More \rightarrow Preselector: Off^1$
AMPT $\rightarrow$ RF Atten Manual: Select the lowest possible setting without overloading <sup>2</sup>
AMPT $\rightarrow$ Ref Level: Set the reference level so that the entire signal is clearly visible; if necessary, go to AMPT $\rightarrow$ Range Log and change the grid scale
TV/radio analyzer/receiver mode
SETUP→Reference Ext: Use the external 10 MHz reference frequency
MODE→TV/Radio Analyzer/Receiver→Digital TV
AMDT Mars Dresslaster Off

 $AMPT \rightarrow More \rightarrow Preselector: Off^{1}$ 

MEAS→Digital TV Settings • TV Standard: OFDM T-DMB/DAB

FREQ→Channel RF: Set based on the transmit frequency

<sup>&</sup>lt;sup>1</sup> Only if a preselector is provided in the instrument

<sup>&</sup>lt;sup>2</sup> Overload warnings appear centered at the top of the display as "IFovl" or "Ovld"

Power

## 3 Measurements

### 3.1 Power

### 3.1.1 Transmitter Output Level

The average power of the 1.536 MHz DAB signal is not dependent on the signal contents, but is rather constant. The DAB channel itself was originally defined at 7/4 MHz = 1.75 MHz as one-fourth of a VHF band III channel. Because the mask filter attenuates the output power between about 0.1 dB and 0.3 dB behind the transmitter output, measurements should be taken before and after the mask filter. Note that as a default, the displayed power includes only the power that is decoupled by the directional coupler. The coupling attenuation can be input using the Ref Level Offset function on the R&S<sup>®</sup>ETL, and is then automatically calculated into the displayed value.

The R&S<sup>®</sup>ETL can measure the signal level directly via the RF input with an accuracy of 1 dB. Use of a separate power sensor allows an accuracy of 0.1 dB to be achieved.

Procedure				
<ul> <li>Perform these steps at the test port:</li> <li>M1, for forward power before the mask filter</li> <li>M2, for reverse power (see Appendix A) before the mask filter</li> <li>M3, for forward power after the mask filter</li> </ul>				
TV/radio analyzer/receiver	Power sensor			
▲ Check that the max. input power is not exceeded; see Section 2.3				
$AMPT {\rightarrow} More {\rightarrow} Ref$ Level Offset: Set to the full coupling attenuation at the test port for immediate compensation				
Feed a signal into the RF input on the R&S <sup>®</sup> ETL (IN1) Connect the power sensor (IN2) (connected to R&S <sup>®</sup> ETL via USB or sensor input) to the test port				
Define the TV/radio analyzer/receiver	MODE→Spectrum Analyzer			
default settings as described in Section 2.4	FREQ→Center: Set to center frequency at mid-channel			
MEAS→Overview→Adjust Attenuation	MENU→Power Meter→Frequency Cou- pling: Center			
	MENU→Power Meter→Power Meter→On			
Read the measured value; see Fig. 2	Read the measured value; see Fig. 3			

•	Att 15 dB				
	ExpLvI 0.00 dBm				
	Level			L.9 dl	Bm
	Ensemble: EN 5503	13 55020	Date & Time(UTC)	06.12.2004,	11:03:13
	Pass	Limit <	Results	< Limit	Unit
	Level	-60.0	1.9	10.0	dBm
	Sideband		Normal		
	Transmission Mode		Mode I, 1536 carriers		
	Carrier Freq Offset	-30000.0	0.1	30000.0	Hz
(t	Bit Rate Offset	-100.0	0.0	100.0	ppm
	MER/EVM (rms)	24.0	33.5		dB
	MER/EVM (peak)	10.0	23.2		dB
im	BER before Viterbi		0.0e-6(3/10)	1.0e-2	
IIII	FIB Errors		0	1	/s
	Subchannel para	meters (Sub	ChId 1, Type STREAM	AUDIO)	
	BER before RS 👝		r maa Hotamioable	2.0e-4	
	Packet Error Ratio	becilic	T-DMB Infobo	1.0e-8	
	Packet Errors ( NO	t in use	for DAB sign	als) 1	/s
	MPEG Ts Bitrate		Not applicable		kbit/s
	9dBm   BER 0.0e-6	L MED DD Ed	B DEMOD	FIC	

Fig. 2: TV/radio analyzer/receiver mode, MEAS $\rightarrow$ Overview menu: The level can be read in the first table row, in the status bar on the test screen or in the zoomed view (MEAS $\rightarrow$ Overview $\rightarrow$ Zoom)

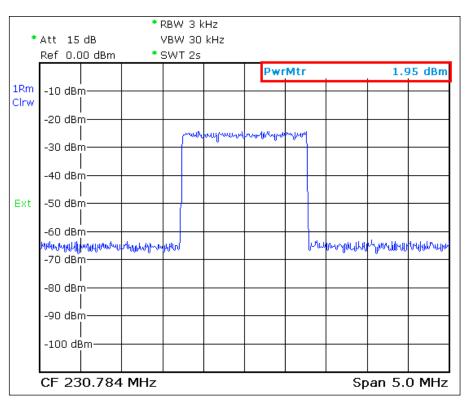


Fig. 3: Spectrum analyzer mode: DAB spectrum with integrated reading from the power sensor displayed at the top right

#### 3.1.2 Crest Factor

It is important to know the crest factor so that the components that follow the transmitter – such as the mask filter, the antenna combiner, the coaxial cable and the antenna – can be adequately dimensioned.

The crest factor (CF) defines the relationship between the highest occurring amplitude of the modulated carrier signal ( $U_{Peak}$ ) and the RMS voltage ( $U_{RMS}$ ) of a signal:

$$CF = 20 \cdot log \frac{U_{Peak}}{U_{RMS}}$$

More recently, however, a new way of defining the crest factor has become prevalent, in which a ratio is formed from the peak envelope power (PEP) and the average power. The resulting value is smaller by an amount equal to the crest factor of the sinus carrier, i.e. 3.01 dB. A crest factor calculated in this way is smaller by an amount equal to the crest factor of the sinus carrier the crest factor of the sinus carrier, i.e. 3.01 dB. [3]

Orthogonal frequency division multiplex (OFDM) signals exhibit a very high crest factor because in extreme cases, all carriers could be overlaid or even eliminated at any given moment. In the case of OFDM, the following equation applies to this theoretical crest factor:

 $Cf_{OFDM} = 10 \cdot \log(2N)$ , where N = number of carriers

Because the signal peaks occur less frequently at high crest factors, any measurement would be valid only for the time period when the measurement was made. This is why the complementary cumulative distribution function (CCDF) includes the statistical probability that a signal peak will occur. The CCDF method determines the peak envelope value, which is why the calculated value must be corrected by a factor of  $\sqrt{2}$ , or 3.01 dB. [2]

DAB provides four different modes (mode I through mode IV), each of which has a different number of subcarriers and a different symbol length. Mode I is used most extensively, with a theoretical crest factor of about 35 dB. In practice, it is limited to about 13 dB in the transmitter.

The mask filter at the transmitter output removes intermodulation products lying outside of the useful band. However, this filtering results in a deformation of the envelope, which then increases the crest factor. This is why, when measuring the crest factor, it is important to distinguish between the crest factor of the transmitter and the crest factor of the bandwidth-limited signal (e.g. after the mask filter). Using the R&S<sup>®</sup>ETL, the crest factor is measured in spectrum analyzer mode. The crest factor of the transmitter is measured directly at the transmitter test port (M1). The crest factor of the bandwidth-limited signal can be measured at the test port after the mask filter (M3).

Procedure: Transmitter crest factor

A Check that the max. input power is not exceeded; see Section 2.3

Connect the R&S<sup>®</sup>ETL (IN1) to the test port before or after the mask filter (M1 / M3)

MODE→Spectrum Analyzer

FREQ→Center: Set to center frequency at mid-channel

AMPT→RF Atten Manual: Select the lowest possible setting without overloading <sup>3</sup>

 $\mathsf{MEAS} \rightarrow \mathsf{More} \rightarrow \mathsf{CCDF} \rightarrow \mathsf{Res} \; \mathsf{BW} : 3 \; \mathsf{MHz}$ 

MEAS→More→CCDF→# of Samples: 1000 000 000

Read crest factor and add 3.01 dB; see Fig. 4

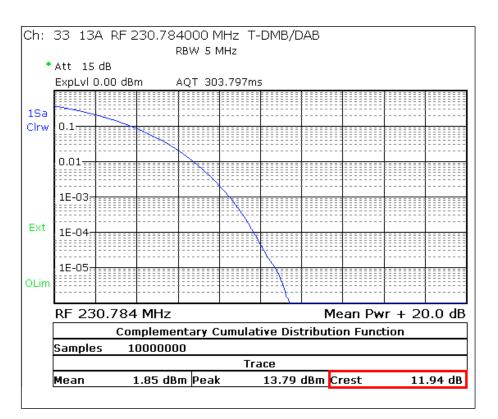


Fig. 4: TV/radio analyzer/receiver mode, MEAS $\rightarrow$ Modulation Analysis $\rightarrow$ CCDF menu: View with the calculated crest factor at the bottom right

<sup>&</sup>lt;sup>3</sup> Overload warnings appear centered at the top of the display as "IFovl" or "Ovld".

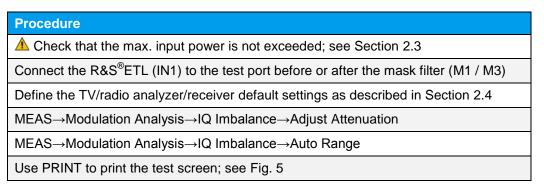
### 3.2 Modulator Characteristics

#### 3.2.1 I/Q Imbalance

DAB modulators are essentially an IFFT signal processing block followed by an I/Q modulator. This I/Q modulator can be either digital or analog. If a DAB modulator uses direct modulation, then the I/Q modulator is analog. In this case, it must be aligned cleanly to minimize the following influencing factors:

- Amplitude imbalance
- Quadrature error
- Carrier suppression

Only a very poor carrier suppression can be detected as a notch directly at mid-band (carrier number 0) on the MER(f). However, because carrier number 0 is not used, it does not cause interference, and it is therefore not visible in the MER. Amplitude imbalance and quadrature error (see Fig. 5) negatively affect the MER of all COFDM carriers. The carriers above the DAB mid-band relate to the carriers under mid-band and vice versa.



#### Modulator Characteristics

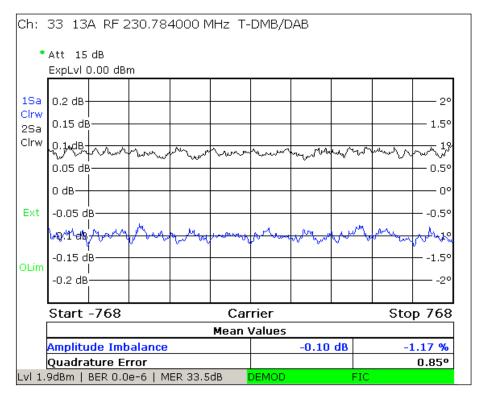


Fig. 5: TV/radio analyzer/receiver mode, MEAS→Modulation Analysis →I/Q Imbalance menu: Detailed analysis of amplitude imbalance and quadrature error over all carriers

#### 3.2.2 Amplitude Frequency Response and Group Delay

In analog televisions, amplitude frequency response and group delay were important parameters for a transmission path between the transmitter output and the receiver input. Because of the differential modulation used with DAB, larger tolerances can now be permitted without noticeable reductions in quality. The mask filter and antenna combiners cause linear distortions. These linear distortions can be compensated by a precorrector within the transmitter. As a result, however, the linear distortions reappear reversed directly at the transmitter output.

Therefore, the preferred method is to measure amplitude frequency response and group delay after all filter stages at a test port in the antenna combiner. Of course, the results will differ at the various measurement points.

#### Procedure

A Check that the max. input power is not exceeded; see Section 2.3

If available, connect the  $R\&S^{\otimes}ETL$  (IN1) to the test port (M4) on the antenna combiner, or else to (M3) after the mask filter

Define the TV/radio analyzer/receiver default settings as described in Section 2.4

MEAS→Channel Analysis→Amplitude & GroupDelay→Auto Range

Use PRINT to print the test screen; see Fig. 6

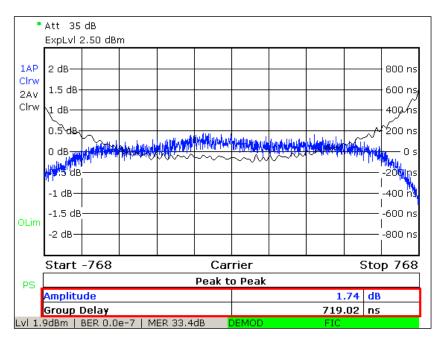


Fig. 6: TV/radio analyzer/receiver mode, MEAS→Channel Analysis→Amplitude & Group Delay menu: Amplitude frequency response and group delay after an uncompensated mask filter

### 3.3 Out-of-Band Emissions

DAB transmitters include very linear AB amplifiers. The transmitted signal is further linearized by a digital precorrection stage in the modulator. In spite of these, some residual nonlinearities remain. These cause intermodulation products to form from the many COFDM carriers.

On the one hand, these additional, unwanted frequency components appear in the channel itself. There, they act as additional disturbance power and therefore reduce the signal quality.

On the other hand, the intermodulation products also occur outside of the channel, and can negatively impact the signal quality of other channels. There are several distinct components:

- Shoulder attenuation Describes the power of the noise components in the near field of the channel boundary
- Adjacent channel emissions
   Components within several MHz of the channel boundaries
- Harmonics
   Components at multiples of the transmit frequency

#### 3.3.1 Shoulder Attenuation and Adjacent Channel Emissions

The mask filter is used to reduce these unwanted out-of-band emissions. Critical mask filters are used when an adjacent channel requires protection, making more stringent requirements for attenuation of out-of-band emissions necessary. All other mask filters are uncritical.

ETSI EN 302077 defines the following three transmitter power classes:

- P < 25 W
- 25 W <= P <= 1000 W
- P > 1000 W

Masks for the following four scenarios are further defined for these four transmitter power classes:

- case 1 (solid line mask VHF T-DAB transmitters operating in critical cases)
- case 2 (dashed line mask VHF T-DAB transmitters operating in uncritical cases or in the L-band)
- case 3 (solid line mask VHF T-DAB transmitters operating in exceptional circumstances to protect safety services)
- case 4 (chain dotted line mask VHF T-DAB transmitters operating in the channel 12D and certain areas)

The high dynamic range of the signal after the mask filter makes it impossible to check adherence to the tolerance mask directly using a spectrum analyzer. This is why an adjustable notch filter is typically used to reduce the useful band power. Before the measurement, the tracking generator on the R&S<sup>®</sup>ETL records the frequency response of the notch filter so that its influence on the measurement results **after the mask filter** can automatically be taken into consideration using the transducer function.

Another option is to use the tracking generator to log the frequency response of the mask filter itself before the measurement so that its influence can be calculated into the spectrum analysis results **before the mask filter** using the transducer function.

Transducer file procedure			
After mask filter using a notch filter	Before mask filter		
Record the frequency response of the adjustable notch filter in a transducer file; see Appendix A	Record the frequency response of the mask filter in a transducer file; see Appendix A		
Connect the R&S <sup>®</sup> ETL TV analyzer (IN1) to the test port after the mask filter (M3)	Connect the R&S <sup>®</sup> ETL TV analyzer (IN1) to the test port before the mask filter (M1)		

The shoulder attenuation as well as the emissions within several MHz of the channel can be measured on the R&S<sup>®</sup>ETL by means of cursor measurements in spectrum analyzer mode.

The out-of-band emission function is a convenient way to comply with all masks defined in the ETSI EN 302077 standard.

Procedure				
Cursor measurement	Out-of-band emission function			
A Check that the max. input power is not exceeded; see Section 2.3				
Go to SETUP $ ightarrow$ Transducer to enable the previously generated transducer file				
Define the spectrum analyzer default set- tings as described in Section 2.4	Define the TV/radio analyzer/receiver de- fault settings as described in Section 2.4			
SWEEP→Sweeptime Manual: Set to 5 s				
MKR→Marker 1: Set to center	MEAS→Spectrum→OutOfBand Emission			
The following three settings must be re- peated for each defined measurement	Go to MEAS→Spectrum→OutOfBand Emission→Out of Band Emission Setup			
point	Select the power range			
MKR→Marker 2: Set to meas- urement point	Select the classification			
MKR→More→Marker 3: Set to the next measurement point	MEAS→Spectrum→Adjust Attenuation			
Read the marker delta values; see Fig. 7. Use PRINT to gener- ate a printout as needed	Use PRINT to print the results; see Fig. 8			
SETUP→Transducer→Active Off: Disable the transducer file				

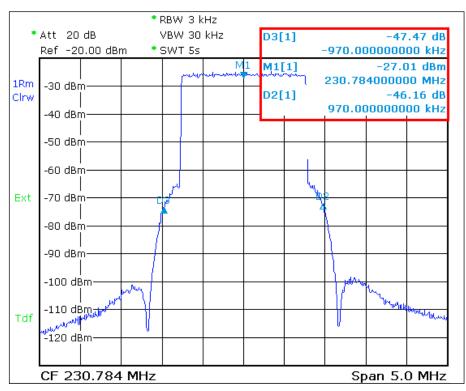


Fig. 7: Spectrum analyzer mode: Measuring the shoulder attenuation using the cursor method with active transducer file at +970 kHz in the 1.536 MHz DAB channel

#### Measurements

#### Out-of-Band Emissions

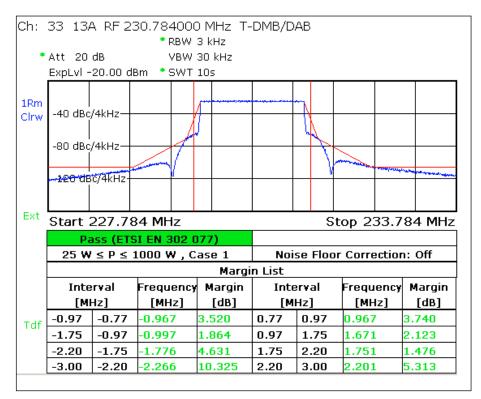


Fig. 8: TV/radio analyzer/receiver mode, MEAS—Spectrum—OutOfBandEmission menu: Measuring the shoulder attenuation and adjacent channel emissions

#### 3.3.2 Harmonics

In addition to adjacent channel emissions, multiples of the transmit frequency can also result in harmonics. A harmonics filter at the transmitter output is used to suppress these harmonics. The R&S<sup>®</sup>ETL TV analyzer can be used to measure out-of-band emissions in spectrum analyzer mode. Because the mask filter does not suppress these harmonics, but rather affects only the channel near range, the harmonics can be measured directly at the test port (M1) on the transmitter output.

The high dynamic range of the signal means that a suitable highpass filter must be used to attenuate the useful channel by at least 40 dB. Notch filters (which are coaxial cavity filters that can be manually adjusted to the channel being suppressed) are not suitable here because they do not attenuate in just the useful band, but rather are repeated at multiples of the useful band. The frequency response of the highpass filter should be documented before the measurement using the tracking generator and then applied during the measurement using the transducer function. The out-of-band components in the L-band cannot be captured using the R&S<sup>®</sup>ETL. Instead, a spectrum analyzer with a 6 GHz span is required.

#### Procedure

A Check that the max. input power is not exceeded; see Section 2.3

Assess the highpass filter and save the result as a transducer file; see Appendix A

Connect the R&S<sup>®</sup>ETL (IN1) to the test port before the mask filter (M1) and add the highpass filter at the auxiliary filter insertion point

Define the spectrum analyzer default settings as described in Section 2.4

FREQ→Center: Set to 1.5 GHz

SPAN→Span Manual: Set to 3 GHz

Go to SETUP  $\!$  Transducer to enable the previously generated transducer file for the highpass filter

Go to MKR $\rightarrow$ Marker 1 and use the marker functions to study the range around the multiples of the transmit frequency; see Fig. 9

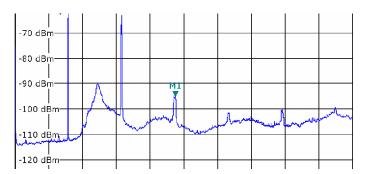


Fig. 9: Spectrum analyzer mode: Useful channel attenuated using the highpass filter; the harmonics, which can be assessed using the marker function, are clearly visible

### 3.4 Signal Quality

#### 3.4.1 Frequency Accuracy

Single-frequency networks (SFN), in particular, place very stringent requirements on the frequency accuracy of a DAB transmitter of less than  $10^{-9}$ . The carrier frequency offset is measured using the R&S<sup>®</sup>ETL in TV/radio analyzer/receiver mode at the test port (M1) of the transmitter output.

Procedure
igtheadrightarrow Check that the max. input power is not exceeded; see Section 2.3
Connect the R&S <sup>®</sup> ETL TV analyzer (IN1) to the test port before the mask filter (M1)
Define the TV/radio analyzer/receiver default settings as described in Section 2.4
Press MEAS→Overview→Adjust Attenuation
Note the carrier frequency offset reading; see Fig. 10

Ch:	33 13A RF 230.	784000 MH	Ηz	T-DMB/DAB			
* Att 15 dB							
ExpLvI 0.00 dBm							
						_	
	Carr Freq Offs	æt			100	.0 m	HZ
	Ensemble: EN 5501	13 55020		Date & Time	(UTC) <mark>0</mark> 6	5.12.2004, :	11:03:06
	Pass	Limit <		Results	<	Limit	Unit
	Level	-60.0			1.9	10.0	dBm
	Sideband			N	Iormal		
	Transmission Mode			Mode I, 1536 ca	arriers		
	Carrier Freq Offset	-30000.0			0.1	30000.0	Hz
Ext	Bit Rate Offset	-100.0			0.0	100.0	ppm
	MER/EVM (rms)	24.0			33.5		dB
	MER/EVM (peak)	10.0			23.1		dB
DLim	BER before Viterbi					1.0e-2	
	FIB Errors				0	1	/s
Subchannel parameters (SubChId 1, Type STREAM AUDIO )							
- 1	BER before RS 👝		- 1	DA DUCATA	iceble .	. 2.0e-4	
- 1	BER before RS Packet Error Ratio	Jecific	-	Not appl	icable	1.0e-8	
- 1	Packet Errors(not	in use	fc	or DAB s	igna	<b>IS)</b> 1	/s
- 1	MPEG Ts Bitrate			Not appl			kbit/s
vl 1.	9dBm   BER 0.0e-6	MER 33.5dl	в	DEMOD		FIC	

Fig. 10: TV/radio analyzer/receiver mode, MEAS $\rightarrow$ Overview menu: The frequency accuracy can be read in the 4th table row, as well as in the zoomed view (MEAS $\rightarrow$ Overview $\rightarrow$ Zoom)

#### 3.4.2 Modulation Error Ratio

The modulation error ratio (MER) is a measure of the sum of all interference that affects a digital TV signal. The deviation of the points in the constellation diagram from their theoretical position is recorded. This makes a quantitative assessment of the signal quality possible. The MER is typically expressed in dB as a logarithmic ratio between the RMS value of the signal amplitude and the error vector magnitude.

 $MER_{RMS} = 20 log_{10} \frac{\sqrt{\frac{1}{N} \sum_{n=0}^{N-1} (|error\_vector|)^2}}{U_{RMS}} [dB]$ 

A high MER value indicates good signal quality. In practice, the MER lies in the range of only a few dB to around 40 dB. A good DAB transmitter has a MER in the range of approximately 33 dB. When receiving DAB signals over a roof antenna with gain, a MER of 20 dB to 30 dB would be measurable at the antenna box. Values between 10 dB and 20 dB are expected for portable receivers with a room antenna.

At the same time, the MER is the single most important quality parameter for a DAB transmitter. The MER can be expressed as an averaged value over all COFDM subcarriers or as MER(f) in a graph via the DAB channel. It is important to note that when used in DAB, differential modulation ( $\pi$ /4-shift-DQPSK) will result in a MER that is about 3 dB worse than that for DVB-T, for example. The following applies to DAB:

$$MER_{RMS}[dB] \leq SNR[dB] - 3dB$$

Procedure
A Check that the max. input power is not exceeded; see Section 2.3
Connect the $R\&S^{@}ETL$ (IN1) to the test port before or after the mask filter (M1 / M3)
Define the TV/radio analyzer/receiver default settings as described in Section 2.4
MEAS→Modulation Analysis→MER(f)→Adjust Attenuation
SPAN→Full Span
Use PRINT to print the test screen; see Fig. 11

High-efficiency transmitters can cause the MER(f) to display a slight distortion after the equalizer.

#### Measurements

Signal Quality

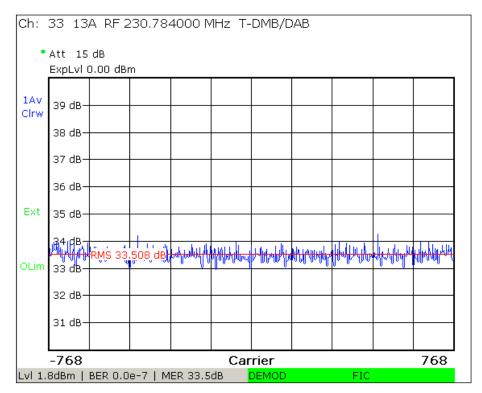


Fig. 11: TV/radio analyzer/receiver mode, MEAS $\rightarrow$ Modulation Analysis $\rightarrow$ MER(f) menu: MER as a function of the frequency and integration of the MER averaged over the channel (RMS).

#### 3.4.3 Constellation Diagram

The constellation diagram makes it possible to display all the signal states that occurred in quadrature modulation at discrete time intervals at the same time. The constellation diagram is a graphical representation of the in-phase and quadrature components of the QAM signal in the x- and y-axes. In the case of modulation with multiple carriers, the constellation diagram typically forms the sum of the signal states of all the carriers. A noisy or disrupted DAB signal will exhibit cloud-like effects. The smaller the resulting points on the constellation diagram, the better the signal quality. When making measurements directly on the transmitter, only fine constellation points should be visible.

Procedure

A Check that the max. input power is not exceeded; see Section 2.3

Connect the R&S<sup>®</sup>ETL (IN1) to the test port before or after the mask filter (M1 / M3)

Define the TV/radio analyzer/receiver default settings as described in Section 2.4

MEAS 

Modulation Analysis 

Const Diagram 

Adjust Attenuation

SPAN→Full Span

Use PRINT to print the constellation diagram; see Fig. 12

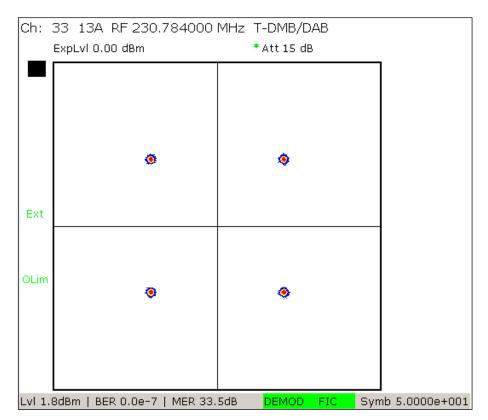


Fig. 12: TV/radio analyzer/receiver mode, MEAS→Modulation Analysis→Const Diagram menu: DAB constellation diagram (layer A, 64QAM)

#### 3.4.4 Bit Error Ratio

The only error correction offered by DAB is convolutional coding that is evaluated using a Viterbi decoder. This method is capable of recognizing and correcting bit errors in the data stream. As a result, the following two bit error ratios (BERs) are available:

- BER before Viterbi
- BER after Viterbi

The DAB frame is split into the fast information channel (FIC) and the main service channel (MSC). The FIC transmits important information for the DAB receiver, while the MSC transmits the actual user data for the subchannels. In DAB, the various subchannels are provided with varying degrees of error correction, i.e. unequal FEC. The FIC has a constant error correction at a code rate of 1/3.

The R&S<sup>®</sup>ETL can be used to measure the BER before Viterbi decoder for the entire channel, for just the FIC, or for just the MSC (BER of all subchannels); see Fig. 13. In the FIC, errors after Viterbi decoder are found in faulty fast information blocks (FIB errors).

Ch:	33-13A RF 230.78	4000 MHz	z T-DMB/DAB		
* Att 15 dB ExpLvl -2.50 dBm					
	Pass	Limit <	Results	< Limit	Unit
	Level	-60.0	1.8	10.0	dBm
	Carrier to Noise	15.0	40.4		dB
	MER (rms)	24.0	33.5		dB
OLim	MER (peak)	10.0	22.7		dB
	EVM (rms)		2.11	4.40	%
	EVM (peak)		7.31	22.00	%
	BER bef. Viterbi		0.0e-7(46/100)	1.0e-2	
	BER bef. Viterbi FIC		0.0e-7(1K03/10K0)	1.0e-2	
	BER bef. Viterbi MSC		0.0e-7(36/100)	1.0e-2	
Lvl 1.	8dBm   BER 0.0e-7   M	ER 33.5dB	DEMOD	FIC	

Fig. 13: TV/radio analyzer/receiver mode, MEAS→Modulation Analysis→Modulation Errors menu: BER before Viterbi

All interference on an DAB transmission path can be expressed as bit error ratios (BER). In the case of a functional DAB transmitter, only the BER before Viterbi can differ from null. It will lie in the range of  $10^{-9}$  or less. With small BERs, it is necessary to select correspondingly long measurement times. For acceptance tests, this will be hours, while it will be minutes for monitoring tests.

Procedure

A Check that the max. input power is not exceeded; see Section 2.3

Connect the R&S $^{\rm \! B}\text{ETL}$  TV analyzer (IN1) to the test port before or after the mask filter (M1 / M3)

Define the TV/radio analyzer/receiver default settings as described in Section 2.4

MEAS→Overview→Adjust Attenuation

Open the MEAS→Measure Log→Configure dialog; see Fig. 14 • Select Enable Measurement Log

- Select the Time Span to define the measurement time
- Trace 1: Select BER before Viterbi FIC
- Trace 2: FIB Errors

 $\mathsf{MEAS}{\rightarrow}\mathsf{Measure}\ \mathsf{Log}{\rightarrow}\mathsf{Clear}$ 

Allow the test - lasting from several minutes to several hours - to run completely

Check the validity of the measurement: The measurement is considered valid if no synchronization loss occurs; see Fig. 15

If the measurement is valid: MEAS → Measure Log → Auto Range

If the measurement is valid: Record the max values or use PRINT to print the results

Configure 🛛 🛛 🛛 🛛				
🗹 Enable Measurement Log				
Time Span	1 minute			
	🔽 Time Span Auto			
Trace 1	BER before Viterbi FIC			
Trace 2	FIB Errors			
🗆 Enable GPS	(external USB GPS device)			

Fig. 14: TV/radio analyzer/receiver mode, MEAS $\rightarrow$ Measure Log $\rightarrow$ Configure menu: Configuration for the BER measurement

OLim	0.002- 1E-03-									2
22.06.2012 19:07:39						22.06.2012 19:08:39				
	Values				Avg	м	in	Max		Current
	BER before Viterbi FIC			С	0.0e	0	0.0e0	0.0	le0	0.0e-6
FIB Errors				0.	0	0.0	1	0.0	0.0	
Lvl 1.9dBm   BER 0.0e-6   MER 33.5dB DEMOD FIC										

Fig. 15: TV/radio analyzer/receiver mode, MEAS→Measure Log menu: BER measurement with the measurement log. Red markers directly above the time axis (here in the 1st and 8th time segments) indicate a loss of synchronization. In this case, the BER measurement is invalid

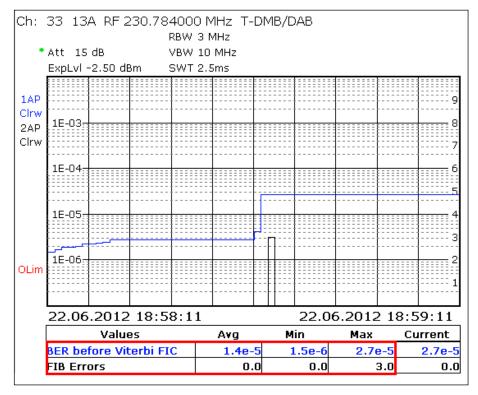


Fig. 16: TV/radio analyzer/receiver mode, MEAS→Measure Log menu: Valid BER measurement

## 4 Abbreviations

BER	Bit error ratio	
CCDF	Complementary cumulative distribution function	
COFDM	Coded orthogonal frequency division multiplex	
DAB	Digital audio broadcasting	
DQPSK	Differential quadrature phase shift keying	
ETI	Ensemble transport interface	
DVB-T	Digital video broadcasting – terrestrial	
MER	Modulation error ratio	
MPEG	Moving Picture Experts Group	
OFDM	Orthogonal frequency division multiplex	
SFN	Single-frequency network	
QAM	Quadrature amplitude modulation	

## 5 References

- [1] "Digital Video and Audio Broadcasting Technology", Walter Fischer, Springer Verlag, 2010, ISBN: 978-3-642-11611-7
- "CCDF determination a comparison of two measurement methods", Christoph Balz, News from Rohde & Schwarz, No. 172 (2001/III), pp. 52 – 53.
- [3] Application Note 7TS02

## 6 Additional Information

Our application notes are regularly revised and updated. Check for any changes at <u>http://www.rohde-schwarz.com</u>.

Please send any comments or suggestions about this application note to Broadcasting-TM-Applications@rohde-schwarz.com.

## 7 Ordering Information

Designation	Туре	Order No.	
Instrument			
TV Analyzer, 500 kHz to 3 GHz, with tracking generator	R&S <sup>®</sup> ETL	2112.0004.13	
Average Power Sensor; 9 kHz to 6 GHz, 200 mW	R&S <sup>®</sup> NRP-Z91	1168.8004.02	
Required options			
One of the following three power sensor interfaces			
- Additional Interfaces	R&S <sup>®</sup> FSL-B5	1300.6108.02	
- Active USB Adapter	R&S <sup>®</sup> NRP-Z3	1146.7005.02	
- Passive USB Adapter	R&S <sup>®</sup> NRP-Z4	1146.8001.02	
Power Sensor Support with NRP	R&S <sup>®</sup> FSL-K9	1301.9530.02	
T-DMB/DAB Firmware	R&S <sup>®</sup> ETL-K250	2112.0533.02	
Measurement Log for DTV	R&S <sup>®</sup> ETL-K208	2112.0579.02	
Recommended options			
Single-frequency network offset			
T-DMB/DAB SFN Frequency Offset	R&S <sup>®</sup> ETL-K251	2112.0540.02	
Picture display			
Video and Audio Hardware Decoder	R&S <sup>®</sup> ETL-B281	2112.0356.02	
HDTV and Dolby Upgrade	R&S <sup>®</sup> ETL-K281	2112.0604.02	
MPEG analysis			
MPEG Analysis/Monitoring	R&S <sup>®</sup> ETL-K282	2112.0610.02	
In-Depth Analysis	R&S <sup>®</sup> ETL-K283	2112.0627.02	
Data Broadcast Analysis	R&S <sup>®</sup> ETL-K284	2112.0633.02	

## A Reverse Power Measurement Uncertainty

Measurement uncertainty occurs during scalar measurements of reverse power as a result of the directivity of measurement couplers. This directivity is an indicator of undesirable forward crosstalk on the reverse power that is being measured. The better the directivity, the less undesirable forward crosstalk is present. A typical directivity value for directional couplers is about -35 dB.

The phase of the overlapping signals must be known in order to measure reverse power exactly. This is possible only with a vector power measurement. However, the scalar measurement offered by the R&S<sup>®</sup>ETLs can also be used to perform the necessary assessment. Instead of determining the precise reverse power value, the R&S<sup>®</sup>ETL ensures that the reverse power is low enough that the transmitter station self-protect function does not shut down the station. This can be determined using a scalar measurement as long as the ratio of the directional coupler directivity to the maximum permissible reverse power is large enough.

During a scalar measurement of the reverse power, the theoretical worst-case measurement errors would be from about +6 dB to  $-\infty$  dB; see Fig. 17. In other words, the reverse power in a scalar measurement can be up to 6 dB too high or else much too low. The measurement uncertainty is dependent on the insertion loss, the directivity, and the reverse power. To simplify the evaluation, the insertion loss should be disregarded because its influence in practice is negligible.

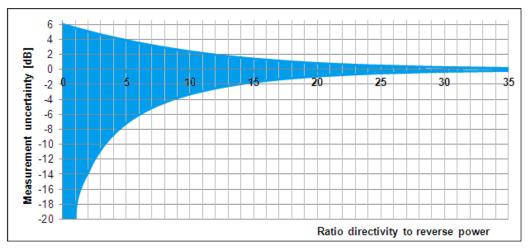


Fig. 17: Measurement uncertainty of the scalar measurement, dependent on the ratio of the directional coupler directivity to the reverse power (insertion loss of the directional coupler is disregarded)

For example, assume that the ratio of the directional coupler directivity to the reverse power is 0 dB (worst case). In this situation, the theoretical maximum measurement error would be between +6 dB and  $-\infty$  dB. However, as long as a 6 dB greater value is acceptable, it is not necessary to determine the actual value.

In another example, assume that the difference between the directional coupler directivity and the reverse power is 20 dB. In this case, the theoretical maximum measurement error would be between 0.83 dB and -0.92 dB. In other words, if the decoupled reverse power is -15 dBm, for example, and the directional coupler directivity is -35 dB, values of between -14.17 dBm and -15.92 dBm can occur at the test instru-

ment. In this case, the measurement uncertainty varies in a range of  $\pm 1$  dB. As a result, a scalar measurement would detect the critical case of a large reverse power.

The following diagram (Fig. 18) can be used to determine the maximum actually reversed power based on the measurement value that is displayed.

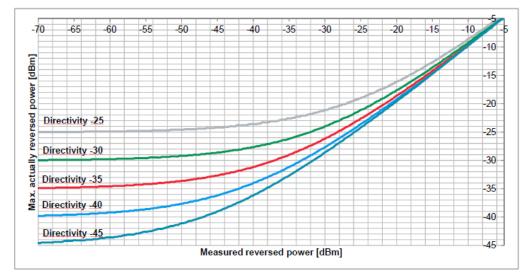


Fig. 18: Maximum actually reversed power based on measured reverse power

In summary, a scalar measurement is sufficient as long as the maximum actually reversed power from the measured line is at an acceptable value.

## B Recording a Filter Frequency Response in a Transducer File

In practice, there are two methods for assessing signals that exceed the dynamic range offered by spectrum analyzers:

- Method 1: The frequency components having the highest power are selectively
  attenuated using auxiliary filters, such as adjustable notch filters or a highpass filter. This reduces the dynamic range enough that the signals can be measured after the auxiliary filter. In order to display the actual dynamic range automatically, a
  transducer file is used to compensate by mathematically subtracting the frequency
  response of the auxiliary filter, which was previously assessed in a separate step.
- Method 2: If the high dynamic range of the signal is achieved by using a specific filter (for example, the mask filter on a transmitter), auxiliary filters are not absolutely required. Instead, the frequency response of the specific filter can be recorded separately as a transducer file. This transducer file is then enabled during the measurement before the filter by adding the filter frequency response, and thus automatically calculating the actual dynamic range.

The transducer file can be created directly using the tracking generator function on the R&S<sup>®</sup>ETL as long as the frequency response of the filter does not exceed the measurable dynamic range<sup>1</sup>:

Generating a transducer file				
MODE→Spectrum Analyzer				
FREQ→Center: Set to center frequency at mid-channel				
SPAN→Span Manual: Set to 10 MHz				
TRACE→Detector Manual Select→More→Detector Average				
BW→Res BW Manual: Set to 30 kHz				
SWEEP→Sweeptime Manual: Set to 2 s				
MENU→Tracking Generator→Source On				
MENU→Tracking Generator→Source Power: Set to 0 dBm				
Connect the cables to be used for the measurement from the Gen Out 50 $\Omega$ output on the R&S <sup>®</sup> ETL to the RF IN 50 $\Omega$ input on the R&S <sup>®</sup> ETL; see Fig. 19:				
AMPT→Ref Level: Set to –30 dBm				
R&S <sup>®</sup> ETL with preselector <sup>2</sup>	R&S <sup>®</sup> ETL without preselector			
AMPT $\rightarrow$ RF Atten Manual: Set to 15 dB	AMPT $\rightarrow$ RF Atten Manual: Set to 0 dB			

<sup>&</sup>lt;sup>1</sup> The frequency response provided in the data sheet can also be entered into the transducer file manually (SETUP→Transducer) <sup>2</sup> If a preselector is provided in the instrument of a preselector is provided in the instrument of a preselector.

<sup>&</sup>lt;sup>2</sup> If a preselector is provided in the instrument, the Preselector setting is available under AMPT $\rightarrow$ More. The preselector is enabled by default

#### Recording a Filter Frequency Response in a Transducer File

#### Generating a transducer file

If an overload occurs<sup>1</sup>, go to AMPT $\rightarrow$ RF Atten Manual and increase the attenuation by 5 dB.

 $MENU \rightarrow Tracking \ Generator \rightarrow Source \ Cal \rightarrow Cal \ Trans$ 

 $\mathsf{MENU}{\rightarrow}\mathsf{Tracking}\;\mathsf{Generator}{\rightarrow}\mathsf{Source}\;\mathsf{Cal}{\rightarrow}\mathsf{Normalize}$ 

Using the previously assessed cables, connect the filter to be assessed from the Gen Out 50  $\Omega$  output on the R&S<sup>®</sup>ETL to the RF IN 50  $\Omega$  input on the R&S<sup>®</sup>ETL; see Fig. 20

Method 1	Method 2			
(reduce the dynamic range using aux-	(assess before increasing the dynamic			
iliary filters)	range)			
MENU→Tracking Generator→Source	MENU→Tracking Generator→Source			
Cal→More→Save As Neg Trd Factor	Cal→More→Save As Pos Trd Factor			
Specify a file name and save the transducer file				
Go to SETUP $\rightarrow$ Transducer $\rightarrow$ Active On to enable the transducer file				



Fig. 19: Connection setup to regulate the cable.



Fig. 20: Connection setup to assess the frequency response of a mask filter.

<sup>1</sup> Overload warnings appear centered at the top of the display as "IFovl" or "Ovld"

#### About Rohde & Schwarz

Rohde & Schwarz is an independent group of companies specializing in electronics. It is a leading supplier of solutions in the fields of test and measurement, broadcasting, radiomonitoring and radiolocation as well as secure communications. Established more than 75 years ago, Rohde & Schwarz has a global presence and a dedicated service network in over 70 countries. Company headquarters are in Munich, Germany.

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- Energy-efficient products
- Continuous improvement in environmental sustainability
- ISO 14001-certified environmental management system



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